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# Massive MIMO-Enabled Semi-Blind Detection for Grant-Free Massive Connectivity

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# Outline

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- **Research Background**
- **System Model**
- **Proposed Semi-Blind Detection Scheme**
- **Simulation Results**



# Background

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## ■ Massive Connectivity

The BS is expected to support wireless connectivity **with billions of devices**

## ■ Grant-Based Random Access

- Require multiple signaling interactions for access scheduling
- Orthogonal multiple access to avoid inter-device interference
- **High access latency, limited number of devices**

## ■ Grant-Free Random Access

- No access scheduling for **reduced access latency**
- Non-orthogonal multiple access for **a larger number of devices**
- Data detection **under severe inter-device interference**

## ■ Training-Based Coherent Detection

- **“Pilot + data”** two phase transmission
- CS-based joint active device detection and channel estimation
- Pilot overhead scales with the number of potential devices
- Rely heavily on accurate CSI



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# System Model

## ■ Grant-Free Massive Connectivity

- BS is equipped with  $N$ -antenna uniform linear array
- Serve  $K$  single-antenna devices,  $K$  is large
- $K_a$  out of  $K$  devices are active,  $K_a$  is far smaller than  $K$
- Communicate with the BS in a grant-free fashion
- Channel and activity remain unchanged during the frame

## ■ Received Signal over $T$ Time Slots

$$\mathbf{R} = \sum_{k=1}^K \alpha_k \tilde{\mathbf{h}}_k \mathbf{x}_k^T + \mathbf{W} = \tilde{\mathbf{H}}\mathbf{X} + \mathbf{W} \quad (1)$$

- $\alpha_k$  is the device activity indicator
- $\tilde{\mathbf{H}} = [\alpha_1 \tilde{\mathbf{h}}_1, \alpha_2 \tilde{\mathbf{h}}_2, \dots, \alpha_K \tilde{\mathbf{h}}_K] \in \mathbb{C}^{N \times K}$  is the massive access channel matrix
- $\mathbf{X} = [\mathbf{x}_1, \mathbf{x}_2, \dots, \mathbf{x}_K]^T \in \mathbb{C}^{K \times T}$  is the access signal matrix



# System Model

## ■ Angular-Domain Received Signal

- To leverage the clustered sparsity of angular-domain channel matrix

$$\mathbf{Y} = \mathbf{A}_R \mathbf{R} = \mathbf{H} \mathbf{X} + \mathbf{N} \quad (2)$$

- Our goal: Jointly infer channel matrix  $\mathbf{H}$  and signal matrix  $\mathbf{X}$  from the overlapped received signal  $\mathbf{Y}$

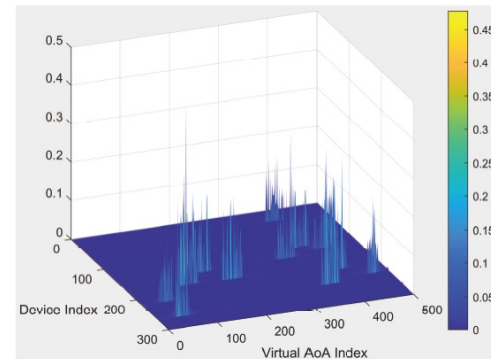


Fig. 1 Clustered sparsity of  $\mathbf{H}$

## ■ Phase and Permutation Ambiguities

- $\Sigma$  is a phase shift matrix,  $\Pi$  is a permutation matrix
- if  $(\hat{\mathbf{H}}, \hat{\mathbf{X}})$  is a solution to the matrix factorization based on (2)
- $(\hat{\mathbf{H}}\Sigma^{-1}\Pi^{-1}, \Pi\Sigma\hat{\mathbf{X}})$  is also a valid solution
- Estimation error  $\|\mathbf{Y} - \hat{\mathbf{H}}\hat{\mathbf{X}}\|_F^2$  is invariant to any phase shifts and permutations



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# Transmitter Design

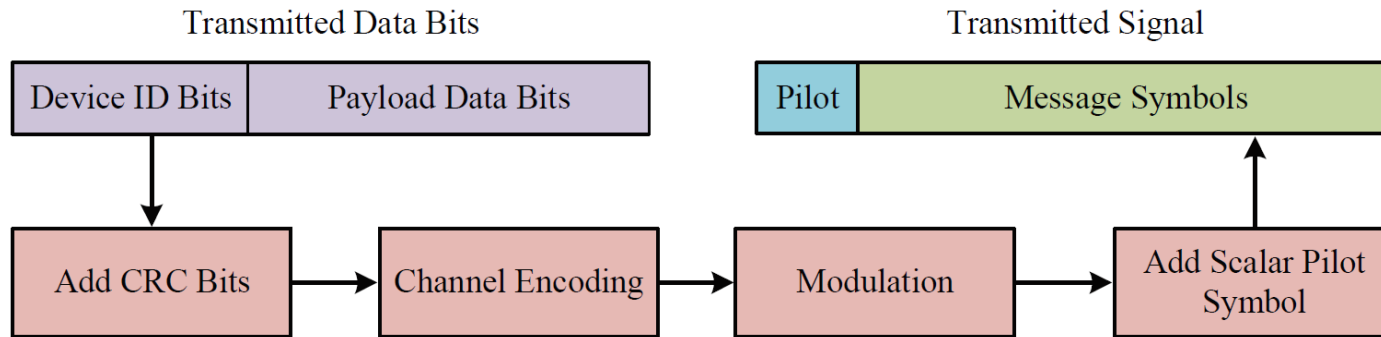


Fig. 1 Transmitter design for semi-blind detection

- A device ID sequence with  $L_i = \lceil \log_2(K) \rceil$  is inserted to **identify  $K$  devices**
- Device ID is the binary representation of the decimal device index
- A  $L_c$ -bit **CRC code** is added to **verify the correctness of the detected data bits**
- Channel coding for error detection and correction
- A scalar pilot symbol is inserted to **eliminate phase ambiguity**





# SIC-Based Semi-Blind Detection

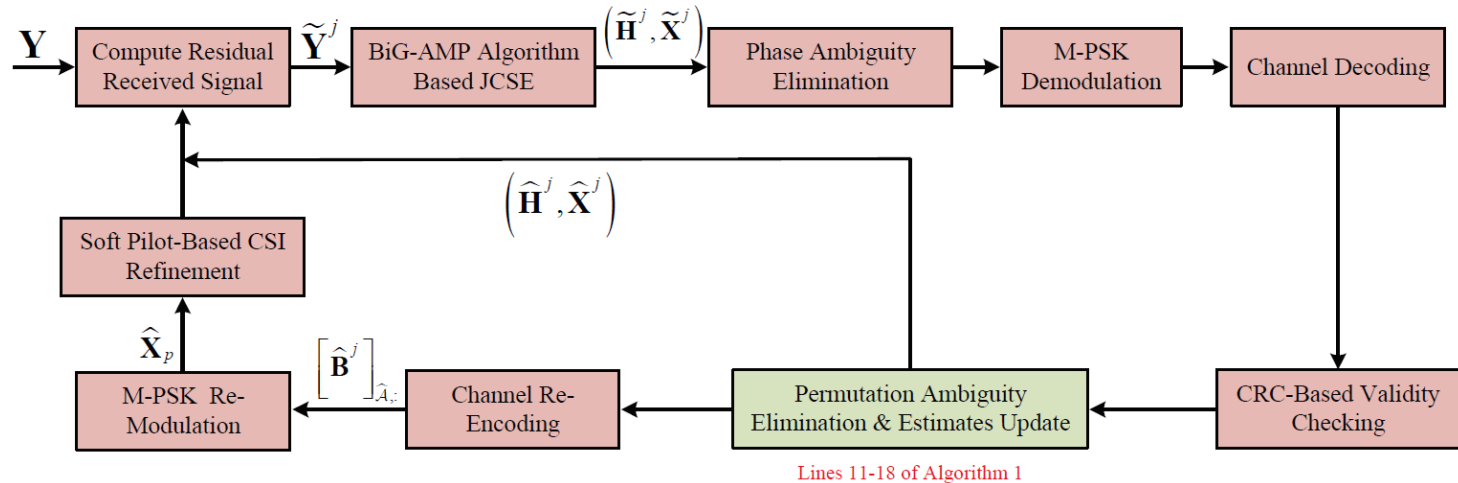


Fig. 3 Block diagram of SIC-based semi-blind detection

## ■ Compute Residual

- Remove the signal components of the detected devices
- Reduce **inter-device interference**

## ■ BiG-AMP-Based JCSE

- Jointly estimate channel and signal matrices
- Leverage the **clustered sparsity** for improved performance



# SIC-Based Semi-Blind Detection

## ■ Eliminate **Phase** Ambiguity

- Each device's phase shift would be the same for all transmitted data symbols

$$\hat{\Sigma} = \text{diag} \left( x_p / \left[ \tilde{\mathbf{X}}^j \right]_{:,1} \right), \tilde{\mathbf{H}}^j = \tilde{\mathbf{H}}^j \hat{\Sigma}^{-1}, \tilde{\mathbf{X}}^j = \hat{\Sigma} \tilde{\mathbf{X}}^j \quad (3)$$

## ■ Demodulation and Decoding

- Obtain the estimated binary data matrix  $\tilde{\mathbf{B}}^j$

## ■ Eliminate **Permutation** Ambiguity

```
for  $k = 1, \dots, K_a^j$  do
   $\hat{k} = \text{bin2dec} \left( \left[ \tilde{\mathbf{B}}^j \right]_{k, \mathcal{I}_{id}} \right), \mathcal{I}_{id} = \{1, \dots, L_i\}$ 
  if  $0 \leq \hat{k} \leq K$  &&  $\mathbf{c}^j(k) == 0$  then
     $\hat{\mathcal{A}}^j = \hat{\mathcal{A}}^{j-1} \cup \hat{k}, \left[ \hat{\mathbf{B}}^j \right]_{\hat{k},:} = \left[ \tilde{\mathbf{B}}^j \right]_{k,:}$ 
     $\left[ \hat{\mathbf{H}}^j \right]_{:,\hat{k}} = \left[ \tilde{\mathbf{H}}^j \right]_{:,k}, \left[ \hat{\mathbf{X}}^j \right]_{\hat{k},:} = \left[ \tilde{\mathbf{X}}^j \right]_{k,:}$ 
  end if
end for
```



# BiG-AMP-Based JCSE Algorithm

## ■ Bayesian inference for exploiting the statistical information

- MMSE estimates of residual channel and signal matrices

$$\left(\tilde{\mathbf{H}}, \tilde{\mathbf{X}}\right) = \iint \frac{1}{Z} p\left(\tilde{\mathbf{Y}}|\mathbf{H}, \mathbf{X}\right) p(\mathbf{H}) p(\mathbf{X}) d\mathbf{H} d\mathbf{X} \quad (4)$$

- Likelihood function

$$p\left(\tilde{\mathbf{Y}}|\mathbf{H}, \mathbf{X}\right) = \prod_{n=1}^N \prod_{t=1}^T \frac{1}{\pi\sigma^2} \exp\left(-\frac{1}{\sigma^2} |\tilde{y}_{n,t} - z_{n,t}|^2\right) \quad (5)$$

- Bernoulli-Gaussian a priori for channel matrix

$$p(\mathbf{H}) = \prod_{n=1}^N \prod_{k=1}^K [(1 - \gamma_{n,k}) \delta(h_{n,k}) + \gamma_{n,k} f(h_{n,k})] \quad (6)$$

- Discrete a priori for signal matrix

$$p(\mathbf{X}) = \prod_{k=1}^K \prod_{t=1}^T \frac{1}{M} \sum_{m=1}^M \delta(x_{k,t} - s_m) \quad (7)$$

- Challenges: high-dimensional integrals, unacceptable complexity



# BiG-AMP-Based JCSE Algorithm

## ■ BiG-AMP Algorithm for Low-Complexity Approximation

- Central-limit theorem and Taylor-series approximation
- The matrix estimation problem is **decoupled into multiple independent scalar estimation problems**
- The marginal posterior distribution of channel elements

$$\hat{q}_{n,k} = h_{n,k} + w_{n,k}^q, \forall n, k, \text{ with } w_{n,k}^q \sim \mathcal{CN}(w_{n,k}^q; 0, v_{n,k}^q) \quad (8)$$



$$\begin{aligned} \forall n, k: v_{n,k}^h(u+1) &= \mathbb{V} \left[ h_{n,k} | \hat{q}_{n,k}(u), v_{n,k}^q(u) \right] \\ \forall n, k: \hat{h}_{n,k}(u+1) &= \mathbb{E} \left[ h_{n,k} | \hat{q}_{n,k}(u), v_{n,k}^q(u) \right] \end{aligned} \quad (9)$$

- The marginal posterior distribution of signal elements

$$\hat{r}_{k,t} = x_{k,t} + w_{k,t}^x, \forall k, t, \text{ with } w_{k,t}^x \sim \mathcal{CN}(w_{k,t}^x; 0, v_{k,t}^r)$$



$$\begin{aligned} \forall k, t: v_{k,t}^x(u+1) &= \mathbb{V} \left[ x_{k,t} | \hat{r}_{k,t}(u); v_{k,t}^r(u) \right] \\ \forall k, t: \hat{x}_{k,t}(u+1) &= \mathbb{E} \left[ x_{k,t} | \hat{r}_{k,t}(u); v_{k,t}^r(u) \right] \end{aligned}$$



# BiG-AMP-Based JCSE Algorithm

## ■ Clustered Sparsity-Based A Priori Refining Strategy

- Employ EM algorithm to learn the unknown hyper-parameters

$$\xi = \{ \sigma^2, \mu, \tau, \gamma_{n,k}, \forall n, k \}$$

- Introduce a constraint term to leverage the clustered sparsity

Traditional E-step of EM algorithm

$$\hat{\xi}(u+1) = \arg \max_{\xi} \left\{ \mathbb{E} \left[ \log p(\mathbf{H}, \mathbf{X}, \mathbf{Z}, \mathbf{Y}; \xi) \mid \mathbf{Y}; \hat{\xi}(u) \right] - \omega \sum_{n=1}^{N_{BS}} \sum_{k=1}^K \sum_{(n',k') \in \mathcal{N}_{n,k}} (\gamma_{n,k} - \gamma_{n',k'})^2 \right\},$$

constrain term



- one element will be nonzero (zero) with a high probability if most of its neighboring elements are non-zero (zero)



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# Numerical Results

## ■ Simulation Setup

- The BS is equipped with  $N = 320$  antennas
- Serve  $K = 500$  single-antenna devices
- The number of active devices  $K_a$  varies from 30 to 80

## ■ Baseline Scheme

- Training-based coherent detection
- GMMV-AMP-based JADCE<sup>[Ke'20]</sup>
- Pilot overhead

$$T_p = \underbrace{(L_i)}_{\text{device ID}} + \underbrace{(L_c)}_{\text{CRC}} + \underbrace{1}_{\text{scalar pilot}} \left\lceil \log_2(M)R \right\rceil + 1$$

Parameter	Value
Radius of BS coverage	2 km
Length of CRC bits $L_c$	8
Length of data bits $L_d$	100
Generator polynomial of CRC $\mathbf{p}_c$	[1, 1, 1, 0, 1, 0, 1, 0, 1]
Carrier frequency	2 GHz
System bandwidth	10 MHz
Number of MPCs $L_k$	$\mathcal{U}(L_k; 30, 60)$
Angular spread in degree	$10^\circ$
Complex gain of MPCs $\beta_{k,l}$	$\mathcal{CN}(\beta_{k,l}; 0, 1)$
Modulation Type	BPSK
Channel coding	(2,1,7) convolutional code
Number of SIC iterations $J$	1 or 5
Termination threshold $\epsilon_{\text{sic}}$	$10^{-5}$

## ■ Performance Metrics

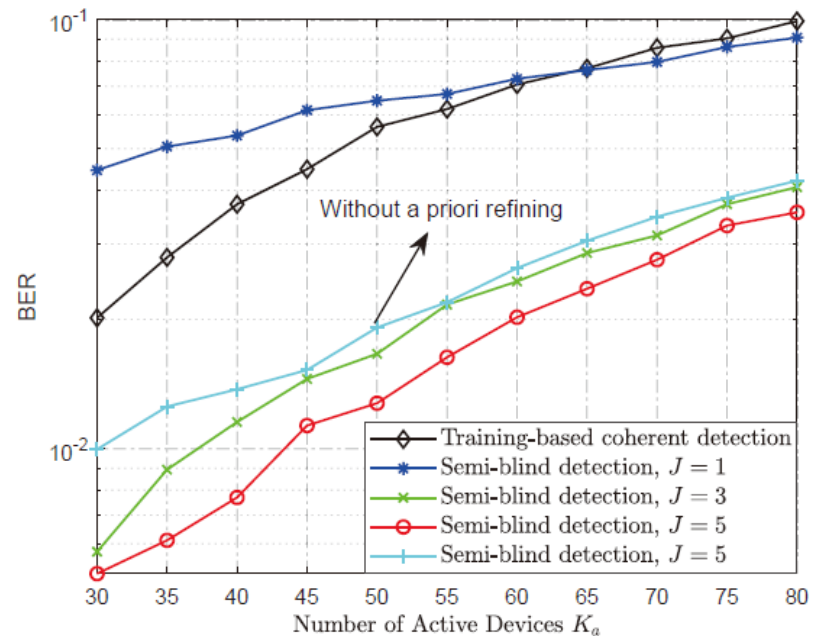
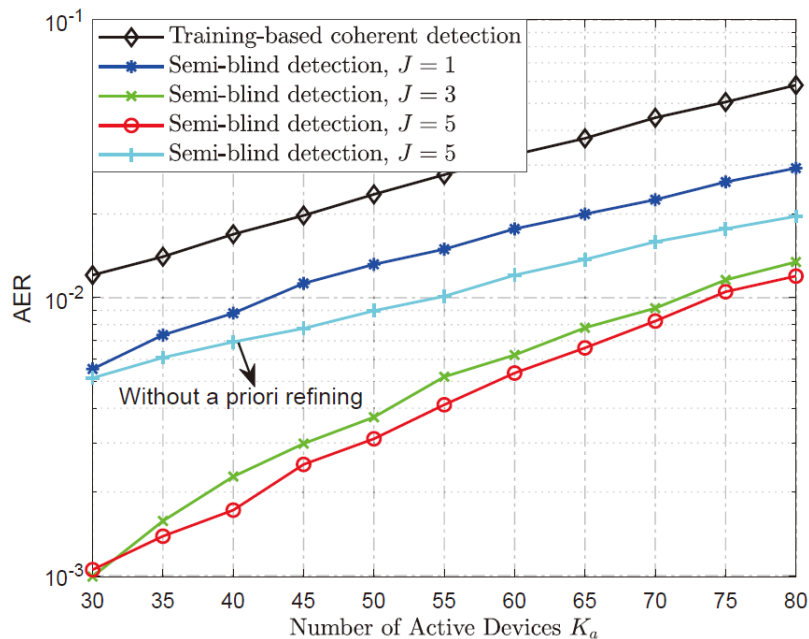
$$\text{AER} = \left| \left( \hat{\mathcal{A}} - \mathcal{A} \right) \cup \left( \mathcal{A} - \hat{\mathcal{A}} \right) \right|_c / K \quad \text{and} \quad \text{BER} = \sum_{k \in \mathcal{A}} \sum_{l \in \mathcal{L}_d} \left| \hat{b}_{k,l} - b_{k,l} \right| / (K_a L_d)$$

[Ke'20] M. Ke et al., "Compressive sensing-based adaptive active user detection and channel estimation: Massive access meets massive MIMO," IEEE Trans. Signal Process., vol. 68, pp. 764-779, 2020.



# Numerical Results

## Simulation Results



- The proposed scheme outperforms the training-based coherent detection scheme
- The performance becomes better as the number of SIC iterations  $J$  increases
- The **clustered sparsity-based** a priori refining further enhances the performance





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# THANKS!

## Q&A



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**ZTE中兴**



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